

# Efficient Algorithm of Well-Localized Bases Construction

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# Agenda

1. Time-Frequency Localization
2. Localization Dilemma
3. Generalized Weyl-Heisenber bases
4. Construction and optimization ideas
5. Another way and orthogonality conditions
6. Winner basis and finally “fast” algorithm
7. OFDM and OFTDM application

# Time-Frequency Localization

The world of transients is considerably larger and more complex  
Than the garden of stationary signals.

Stephane Mallat

## Fourier transform

- Linear time invariant systems

$$\forall \omega \in \mathbb{R}, L e^{i\omega t} = \hat{\lambda}(\omega) e^{i\omega t}$$

$$L f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \hat{\lambda}(\omega) e^{i\omega t} d\omega, \quad \hat{\lambda}(\omega) = \int_{-\infty}^{\infty} f(t) e^{-i\omega t} dt$$

## Window Fourier transform

- Gabor atoms

$$g_{u,\xi}(t) = g(t-u) e^{i\xi t},$$

- $Sf(u,\xi) = \int_{-\infty}^{\infty} f(t) g_{u,\xi}^*(t) dt = \int_{-\infty}^{\infty} f(t) g(t-u) e^{-i\xi t} dt.$

# Localization Dilemma

## Heisenberg uncertainty relation

$$\left( \int_{-\infty}^{\infty} (t - \tau)^2 |g(t)|^2 dt \right) \left( \int_{-\infty}^{\infty} (f - \nu)^2 |G(f)|^2 df \right) \geq \frac{\|g(t)\|^4}{16\pi^2}$$

## Balian-Low theorem

- It is impossible to construct orthogonal well-localized basis in Hilbertian spaces with ordinary scalar product without the loss in spectral efficiency

# Dilemma Decision

1. Utilization of several initializing functions
2. Hilbertian spaces with real scalar product

$$\langle \varphi_k(t-qT), \varphi_i(t-pT) \rangle \triangleq \text{Re} \left\{ \int_{-\infty}^{\infty} \varphi_k(t-qT) \varphi_i^*(t-pT) dt \right\} = \delta[k-i] \delta[p-q]$$

# Generalized Weyl-Heisenberg basis

$$\mathcal{B} = \{ \psi_{k,l}(t) = g(t - l\tau_0) \exp(j\varphi_{l,k}) \exp(2\pi jk\nu_0 t) \},$$

$$k = 0, \dots, M-1; l = 0, \pm 1, \pm 2, \dots$$

$$s \ n = \sum_{k=0}^{M-1} \left( \sum_{l=0}^{L-1} c_{k,l}^R \psi_{k,l}^R \ n + \sum_{l=0}^{L-1} c_{k,l}^I \psi_{k,l}^I \ n \right), \ n \in J_N \quad \tau_0 \nu_0 = 1/2$$

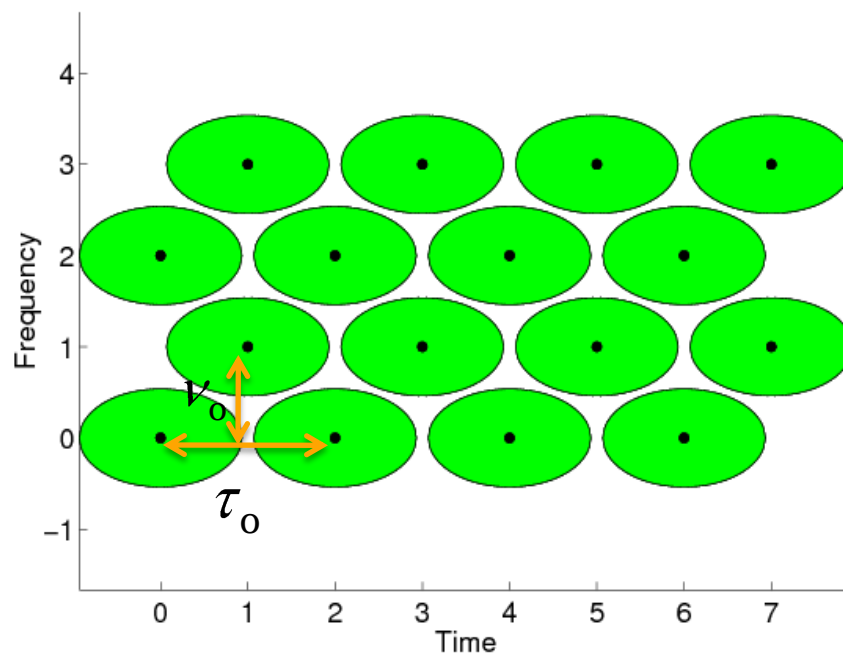
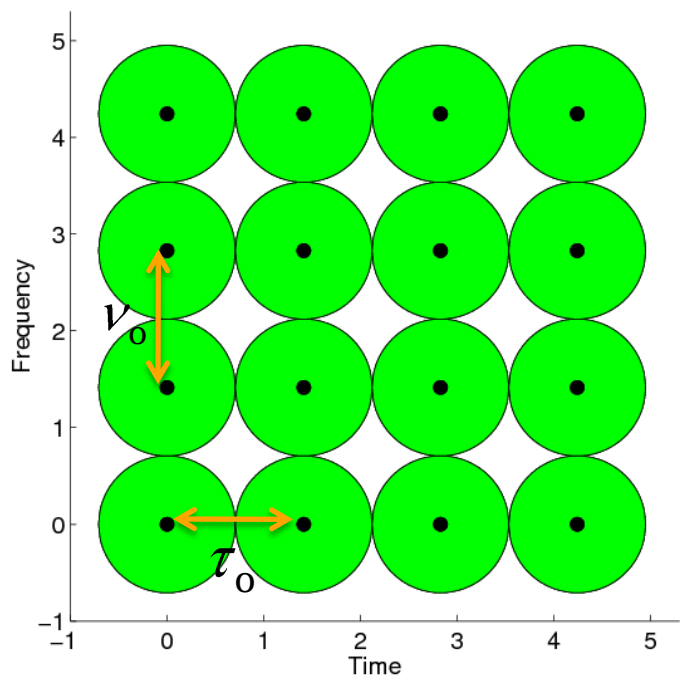
$$\psi_{k,l}^R \ n = g \left[ n - lM \mod N \right] \exp \left( j \frac{2\pi}{M} k \ n - \alpha/2 \right),$$

$$\psi_{k,l}^I \ n = -jg \left[ n + M/2 - lM \mod N \right] \exp \left( j \frac{2\pi}{M} k \ n - \alpha/2 \right),$$

$$\mathbf{U} = \mathbf{U}_R, \mathbf{U}_I$$

$$\mathcal{B} \ J_N \triangleq \psi_{k,l}^R \ n, \psi_{k,l}^I \ n,$$

# Basic functions packing in time-frequency lattice



# Optimal W-G basis (general approach)

Gaussian initializing impulse with necessary parameters

Symmetrical N-periodical approximation

WG-basis parameters initialization:  $M, L, \alpha$

Optimal initializing impulse construction (for WG basis)

Optimal WG-basis parameters search

Orthogonal matrix of WG basis

Fast signal filtering algorithm

$$g_G(t) \longrightarrow g_G(n)$$

$$g_G(n) \longrightarrow \tilde{g}_G(n)$$

$$g_o(n): \min_{g \in \mathcal{U}} \|\tilde{g}_G - g\|_E^2 \triangleq \varepsilon$$

+

Orthogonality conditions

$$\hat{\alpha}, \hat{M}, \hat{L}: \min_{\alpha, M, L} \varepsilon$$

$$\mathbf{U} = \{g_o^{(n,n)}\}_{n=1}^N, N = KL$$

Basis matrix  $\mathbf{U}$  as the multiplication of sparse matrixes



## Construction Ideas

Closeness to the Gabor basis (Gaussian initializing functions)

$$\left\{ \begin{array}{l} \operatorname{Re}(\mathbf{U}^* \mathbf{U}) = \mathbf{I}_{2N} \\ \mathbf{U}_{onm} : \min_{U \in \mathfrak{A}} \|\mathbf{G} - \mathbf{U}\|_E^2, \end{array} \right.$$

To derive orthogonality conditions directly for the initializing function

$$\sum_{r=0}^{2L-1} g\left[\left(n - rM/2\right)_N\right] g\left[\left(n - rM/2 - lM\right)_N\right] = \frac{2}{M} \delta_{l,0}, \forall n \in \{0, \dots, N-1\}$$

# Winner Transform

- Winner Transform

$$\eta_k n = \sum_{r=0}^{2L-1} g \left[ n - r \frac{N}{2L} \right] \exp \left( \frac{2\pi j}{2L} rk \right).$$

- Inverse Winner transform

$$g n = \frac{1}{2L} \sum_{k=0}^{2L-1} \eta_k n .$$

- WG basis orthogonality condition

$$\left| \eta_k^{M/2} n \right|^2 + \left| \eta_{k+L}^{M/2} n \right|^2 = 4/M .$$

# Fast construction algorithm

$$\eta_k^{M/2} n = \frac{2\tilde{\eta}_k^{M/2} n}{\sqrt{M|\tilde{\eta}_k^{M/2} n|^2 + M|\tilde{\eta}_{k-L}^{M/2} n|^2}}, \quad \tilde{\eta}_k^{M/2} n = \sum_{r=0}^{2L-1} g_o \left[ n - rM/2_N \right] \exp\left(\frac{2\pi j}{2L} rk\right)$$

Construct matrix  $\mathbf{Z}$  from the vector of shifts of  $g_0$  as columns  $\tilde{g}_0 n \triangleq g_0 n, g_0[n - N/2L_N], \dots, g_0[n - (2L-1)N/2L_N]^T$

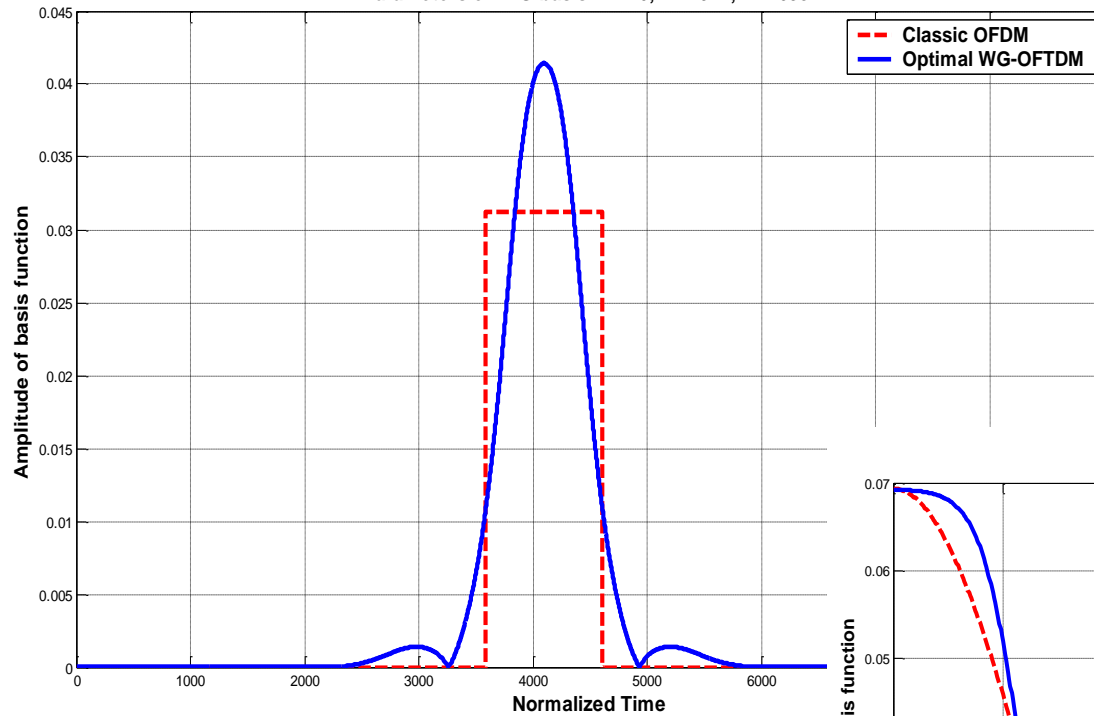
Winner transform on  $g_0$   
 $\leftrightarrow$  FFT on matrix  $\mathbf{Z}$  rows

Construct Winner basis  $\eta_k^{M/2} n$

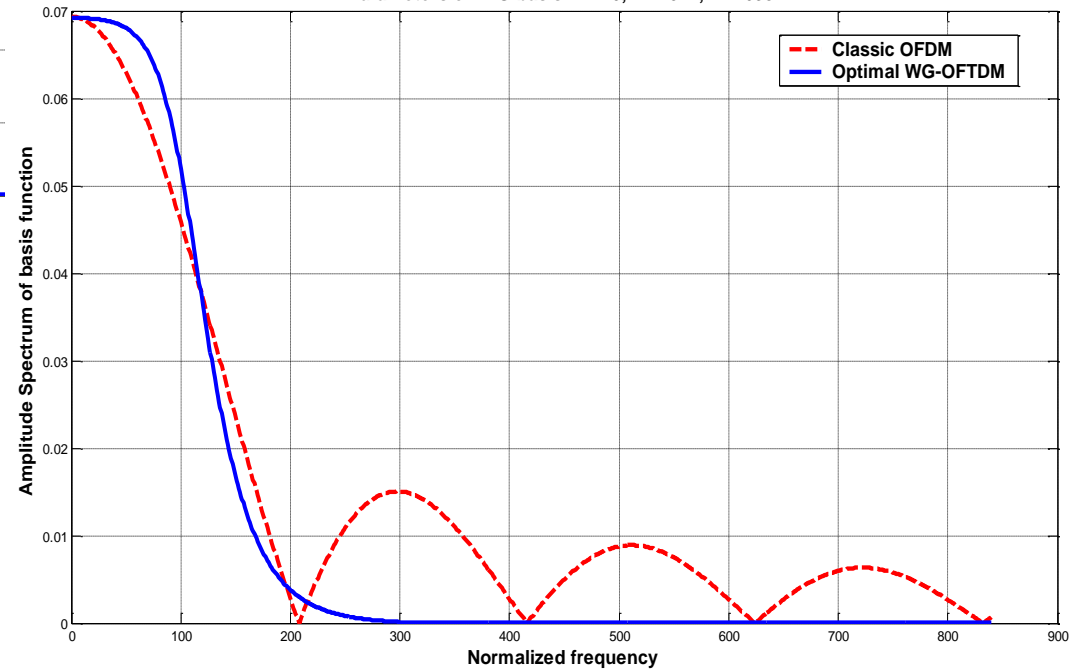
Inverse Winner transform

# Modeling results

Parameters of WG-basis : L=16; M=1024; N=16384



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# OFDM and OFTDM

## Advantages

*No complex equalization and ISI robustness*

*High computational efficiency*

*High bit rate*

## Disadvantages

*Reduced spectral efficiency (Cyclic Prefix)*

*Frequency offset and Doppler shifts sensibility*

*Low frequency localization*

## Disadvantages

*More complex computational algorithm*

## Advantages

*Lower Doppler shift sensibility*

*Lower off-frequency emission*

*Higher spectral efficiency (dense TF packing)*

# Q&A



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